

Single Output

HIGH RELIABILITY DC-DC CONVERTER

 16-40V
 16-50V
 9V
 2.75A
 88% @ 1.37A / 88% @ 2.75A

 Continuous Input
 Transient Input
 Output
 Output
 Efficiency

Full Power Operation: -55°C to +125°C

The MilQor® series of high-reliability DC-DC converters brings SynQor's field proven high-efficiency synchronous rectifier technology to the Military/Aerospace industry. SynQor's innovative QorSeal™ packaging approach ensures survivability in the most hostile environments. Compatible with the industry standard format, these converters operate at a fixed frequency, have no opto-isolators, and follow conservative component derating guidelines. They are designed and manufactured to comply with a wide range of military standards.

Design Process

MQHR series converters are:

- Designed for reliability per NAVSO-P3641-A guidelines
- Designed with components derated per:
 - MIL-HDBK-1547A
 - NAVSO P-3641A

Qualification Process

MQHR series converters are qualified to:

- MIL-STD-810F
 - consistent with RTCA/D0-160E
- SynQor's First Article Qualification
 - consistent with MIL-STD-883F
- SynQor's Long-Term Storage Survivability Qualification
- SynQor's on-going life test

In-Line Manufacturing Process

- AS9100 and ISO 9001:2008 certified facility
- Full component traceability
- Temperature cycling
- Constant acceleration
- •24, 96, 160 hour burn-in
- Three level temperature screening





DESIGNED & MANUFACTURED IN THE USA FEATURING QORSEAL** HI-REL ASSEMBLY

Features

- Fixed switching frequency
- No opto-isolators
- Output over-voltage shutdown
- Remote sense
- Clock synchronization
- Primary referenced enable
- Continuous short circuit and overload protection
- Input under-voltage and over-voltage shutdown

Specification Compliance

MQHR series converters (with MQHE filter) are designed to meet:

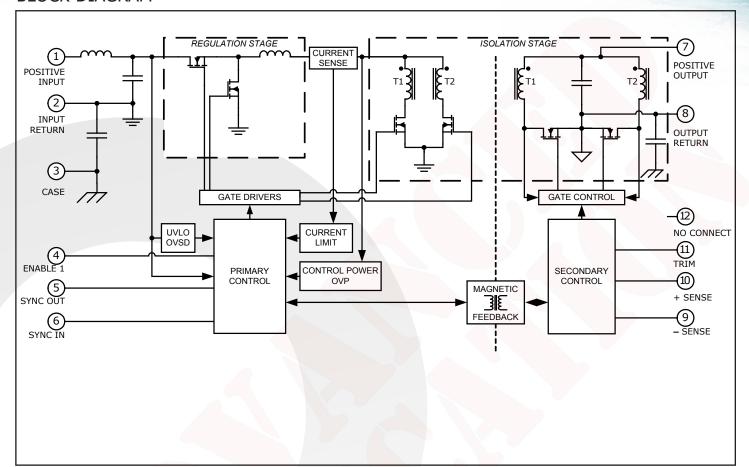
- MIL-HDBK-704-8 (A through F)
- RTCA/DO-160 Section 16, 17, 18
- MIL-STD-1275 (B, D) for V_{IN} > 16V
- DEF-STAN 61-5 (part 6)/(5, 6) for V_{IN} > 16V
- MIL-STD-461 (C, D, E, F)
- RTCA/DO-160(E, F, G) Section 22



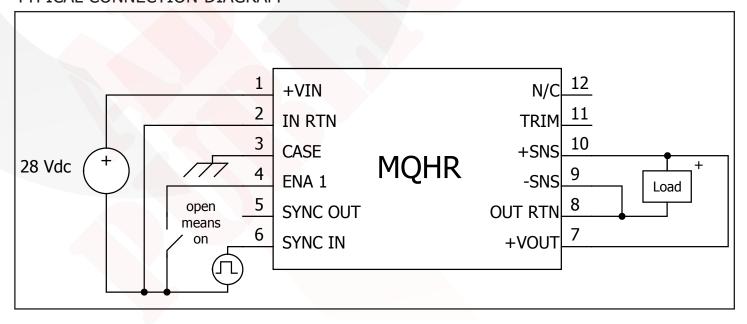
MQHR-28-09S Output: 9V

Current: 2.75A

BLOCK DIAGRAM



TYPICAL CONNECTION DIAGRAM





Output: 9V

Current: 2.75A

MOHR	-28-09S	ELECTRICAL	CHARACTERISTI	CS
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MQHR-28-09S ELECTRICAL CHARACTERISTICS							
Parameter	Min.	Тур.	Max.	Units	Notes & Conditions Vin=28V dc ±5%, Iout=2.75A, CL=0µF, free running	Group A Subgroup	
					(see Note 9) unless otherwise specified	(see Note 11)	
ABSOLUTE MAXIMUM RATINGS							
Input Voltage Non-Operating			60	V			
Operating			60	V	See Note 1		
Reverse Bias (Tcase = 125°C)			-0.8	V	555 1155 2		
Reverse Bias (Tcase = -55°C)			-1.2	V			
Isolation Voltage (I/O to case, I to O)	F00		F00	\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \			
Continuous Transient (≤100µs)	-500 -800		500 800	V			
Operating Case Temperature	-55		125	°C	HB Grade Products, See Notes 2 & 13		
Storage Case Temperature	-65		135	°C			
Lead Temperature (20s)			300	°C			
Voltage at ENA1	-1.2		50	V			
INPUT CHARACTERISTICS Operating Input Voltage Range	16	28	40	V	Continuous	1, 2, 3	
" " " "	16	28	50	V	Transient, 1s	4, 5, 6	
Input Under-Voltage Shutdown					See Note 3		
Turn-On Voltage Threshold	14.75	15.50	16.00	V		1, 2, 3	
Turn-Off Voltage Threshold	14.00	14.75	15.50	V		1, 2, 3 1, 2, 3	
Shutdown Voltage Hysteresis Input Over-Voltage Shutdown	0.65	0.85	1.05	V	See Note 3	1, 2, 3	
Turn-Off Voltage Threshold	52.0	55.0	58.0	V	See Note 3	1, 2, 3	
Turn-On Voltage Threshold	50.5	54.0	56.5	V		1, 2, 3	
Shutdown Voltage Hysteresis	1.0	2.0	3.0	V		1, 2, 3	
Maximum Input Current		0	2	MA	Vin = 16V; Iout = 2.75A	1, 2, 3 1, 2, 3	
No Load Input Current (operating) Disabled Input Current		10	15	mA	Vin = 16V, 28V, 50V; ENA	1, 2, 3	
Input Terminal Current Ripple (pk-pk)		40	75	mA	Bandwidth = 100kHz – 10MHz; see Figure 14	1, 2, 3	
OUTPUT CHARACTERISTICS							
Output Voltage Set Point (Tcase = 25°C)	8.91	9.00	9.09	V	Vout at sense leads	1	
Output Voltage Set Point Over Temperature Output Voltage Line Regulation	8.85 -35	9.00	9.15	V mV	"; Vin = 16V, 28V, 40V; Iout=2.75A	2, 3 1, 2, 3	
Output Voltage Line Regulation Output Voltage Load Regulation	-35	0	35	mV	"; Vout @ (Iout=0A) - Vout @ (Iout=2.75A)	1, 2, 3	
Total Output Voltage Range	8.82	9.00	9.18	V	" 1000 (2000 01) 1000 (2000 21) 31)	1, 2, 3	
Output Over-Voltage Shutdown	11.2	12.2	14.2	V		See Note 5	
Output Voltage Ripple and Noise Peak to Peak		15	75	mV	Bandwidth = 10MHz; CL=11µF	1, 2, 3	
Operating Output Current Range Operating Output Power Range	0		2.75 25	A W		1, 2, 3 1, 2, 3	
Output DC Current-Limit Inception	2.9	3.4	4	A	See Note 4	1, 2, 3	
Back-Drive Current Limit while Enabled	,	0.7		A		1, 2, 3	
Back-Drive Current Limit while Disabled		10	50	mA		1, 2, 3	
Maximum Output Capacitance			750	μF		See Note 5	
DYNAMIC CHARACTERISTICS Output Voltage Deviation Load Transient					See Note 6		
For a Pos. Step Change in Load Current	-500	-250		mV	Total Iout step = 1.375A to 2.75A, 0.275A to 1.375A;	4, 5, 6	
For a Neg. Step Change in Load Current		250	500	mV	n .	4, 5, 6	
Output Voltage Deviation Line Transient					Vin step = 16V to 50V; CL=11µF; see Note 7		
For a Pos. Step Change in Line Voltage	-125		125	mV		4, 5, 6	
For a Neg. Step Change in Line Voltage Turn-On Transient	-125		125	mV		4, 5, 6	
Output Voltage Rise Time		6	10	ms	Vout = 0.9V to 8.1V; Full Resistive Load	4, 5, 6	
Output Voltage Overshoot		0	2	%	Resistive load	See Note 5	
Turn-On Delay, Rising Vin		5.5	8.0	ms	ENA = 5V; see Notes 8 & 10	4, 5, 6	
Turn-On Delay, Rising ENA		3.0	6.0	ms	See Note 10	4, 5, 6	
Restart Inhibit Time Short Circuit Start Time	12	100	150 20	ms ms	See Note 10 Duration of pulse width, see Figure 16	4, 5, 6	
EFFICIENCY	12	17	20	1115	Paradon of paise width, see Figure 10		
Iout = 2.75 A (16 Vin)	85	89		%		1, 2, 3	
Iout = 1.375 A (16 Vin)	86	89		%		1, 2, 3	
Iout = 2.75 A (28 Vin)	84	88		%		1, 2, 3	
Iout = 1.375 A (28 Vin) Iout = 2.75 A (40 Vin)	85 83	88 87		% %		1, 2, 3 1, 2, 3	
Iout = 1.375 A (40 Vin)	83	86		%		1, 2, 3	
Iout = 2.75 A (50 Vin)	82	86		%		0	
Load Fault Power Dissipation		2.5		W	Sustained short circuit on output		

Product# MQHR-28-09S



Output: 9V

Current: 2.75A

MOHR-28-09S ELECTRICAL CHARACTERISTICS (Continued)

Parameter	Min.	Тур.	Max.	Units	Notes & Conditions Vin=28V dc ±5%, Iout=2.75A, CL=0μF, free running	Group A Subgroup
					(see Note 9) unless otherwise specified	(see Note 11)
ISOLATION CHARACTERISTICS						
Isolation Voltage					Dielectric strength	-
Input RTN to Output RTN	500			V		1
Any Input Pin to Case	500			V		1
Any Output Pin to Case	500			V		1
Isolation Resistance (in rtn to out rtn)	100			MΩ		1
Isolation Resistance (any pin to case)	100			MΩ		1
Isolation Capacitance (in rtn to out rtn)		22		nF		1
FEATURE CHARACTERISTICS						
Switching Frequency (free running)	500	550	600	kHz		1, 2, 3
Synchronization Input						
Frequency Range	500		700	kHz		1, 2, 3
Logic Level High	2.0		5.5	V		1, 2, 3
Logic Level Low	-0.5		0.8	V		1, 2, 3
Duty Cycle	20		80	%		See Note 5
Synchronization Output						
Pull Down Current	20			mA	VSYNC OUT = 0.8V	See Note 5
Duty Cycle	40		60	%	Output connected to SYNC IN of other MQHR unit	See Note 5
Enable Control (ENA)						
Off-State Voltage			0.8	V		1, 2, 3
Module Off Pulldown Current	80			μA	Current drain required to ensure module is off	See Note 5
On-State Voltage	2			V		1, 2, 3
Module On Pin Leakage Current			20	μA	Imax draw from pin allowed with module still on	See Note 5
Pull-Up Voltage	3.2	4.0	4.8	V	See Figure A	1, 2, 3
RELIABILITY CHARACTERISTICS						
Calculated MTBF (MIL-STD-217F2)						
GB @ Tcase = 70°C		4030		10 ³ Hrs.		
AIF @ Tcase = 70°C		338		10 ³ Hrs.		
WEIGHT CHARACTERISTICS			172			
Device Weight		45		g		

Electrical Characteristics Notes

- 1. Converter will undergo input over-voltage shutdown.
- 2. Derate output power to 50% of rated power at Tcase = 135°C. 135°C is above the specified operating range. See Figure 5.
- 3. High or low state of input voltage must persist for about 200µs to be acted on by the shutdown circuitry.
- 4. Current limit inception is defined as the point where the output voltage has dropped to 90% of its nominal value. See Current Limit discussion in Features Description section.
- 5. Parameter not tested but guaranteed to the limit specified.
- 6. Load current transition time ≥ 10µs.
- 7. Line voltage transition time $\geq 100 \mu s$.
- 8. Input voltage rise time ≤ 250µs.
 9. Operating the converter at a synchronization frequency above the free running frequency will cause the converter's efficiency to be slightly reduced and it may also cause a slight reduction in the maximum output current/power available. For more information consult the factory.

 10. After a disable or fault event, module is inhibited from restarting for 100ms. See Shut Down section of the Control Features description.

 11. Only the ES and HB grade products are tested at three temperatures. The C grade products are tested at one temperature. Please refer to the
- Construction and Environmental Stress Screening Options table for details.

 12. These derating curves apply for the ES and HB grade products. The C grade product has a maximum case temperature of 70°C.

 13. The specified operating case temperature for ES grade products is -45°C to 100°C. The specified operating case temperature for C grade products
- is 0°C to 70°C.

BASIC OPERATION AND FEATURES

The MQHR DC/DC converter uses a two-stage power conversion topology. The first, or regulation, stage is a buck-converter that keeps the output voltage constant over variations in line, load, and temperature. The second, or isolation, stage uses transformers to provide the functions of input/output isolation and voltage transformation to achieve the output voltage required.

Both the regulation and the isolation stages switch at a fixed frequency for predictable EMI performance. The isolation stage switches at one half the frequency of the regulation stage, but due to the push-pull nature of this stage it creates a ripple at double its switching frequency. As a result, both the input and the output of the converter have a fundamental ripple frequency of about 550 kHz in the free-running mode.

Rectification of the isolation stage's output is accomplished with synchronous rectifiers. These devices, which are MOSFETs with a very low resistance, dissipate far less energy than would Schottky diodes. This is the primary reason why the MQHR converters have such high efficiency, particularly at low output voltages.

Besides improving efficiency, the synchronous rectifiers permit operation down to zero load current. There is no longer a need for a minimum load, as is typical for converters that use diodes for rectification. The synchronous rectifiers actually permit a negative load current to flow back into the converter's output terminals if the load is a source of short or long term energy. The MQHR converters employ a "backdrive current limit" to keep this negative output terminal current small.

There is a control circuit in the MQHR converter that determines the conduction state of the power switches. It communicates across the isolation barrier through a magnetically coupled device. No opto-isolators are used.

An input under-voltage shutdown feature with hysteresis is provided, as well as an input over-voltage shutdown and an output over-voltage limit. There is also an output current limit that is nearly constant as the load impedance decreases (i.e., there is not fold-back or fold-forward characteristic to the output current under this condition). When a load fault is removed, the output voltage rises exponentially to its nominal value without an overshoot. If a load fault pulls the output voltage below about 60% of nominal, the converter will shut down to attempt to clear the load fault. After a short delay it will try to auto-restart.

The MQHR converter's control circuit does not implement an over-temperature shutdown.

The following sections describe the use and operation of additional control features provided by the MQHR converter.

CONTROL FEATURES

ENABLE: The MQHR converter has one enable pin, ENA1 (pin 4), which is referenced with respect to the converter's input return (pin 2). It must have a logic high level for the converter to be enabled; a logic low inhibits the converter.

The enable pin is internally pulled high so that an open connection will enable the converter. Figure A shows the equivalent circuit looking into the enable pin. It is TTL compatible and has hysteresis.

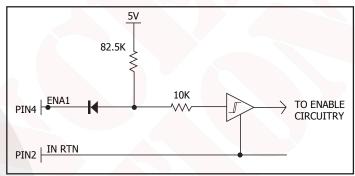


Figure A: Circuit diagram shown for reference only, actual circuit components may differ from values shown for equivalent circuit.

SHUT DOWN: The MQHR converter will shut down in response to only five conditions: ENA input low, VIN input below under-voltage shutdown threshold, VIN input above over-voltage shutdown threshold, output voltage below the output under-voltage threshold, and output voltage above the output over-voltage threshold. Following any shutdown event, there is a startup inhibit delay which will prevent the converter from restarting for approximately 100ms. After the 100ms delay elapses, if the enable inputs are high and the input voltage is within the operating range, the converter will restart. If the VIN input is brought down to nearly 0V and back into the operating range, there is no startup inhibit, and the output voltage will rise according to the "Turn-On Delay, Rising Vin" specification.

REMOTE SENSE: The purpose of the remote sense pins is to correct for the voltage drop along the conductors that connect the converter's output to the load. To achieve this goal, a separate conductor should be used to connect the +SENSE pin (pin 10) directly to the positive terminal of the load, as shown in the connection diagram on Page 2. Similarly, the –SENSE pin (pin 9) should be connected through a separate conductor to the return terminal of the load.

NOTE: Even if remote sensing of the load voltage is not desired, the +SENSE and the -SENSE pins must be connected to +Vout (pin 7) and OUTPUT RETURN (pin 8), respectively, to get proper regulation of the converter's output. If they are left open, the converter will have an output voltage that is approximately 200mV higher than its specified value.

Inside the converter, +SENSE is connected to +Vout with a 100Ω resistor and -SENSE is connected to OUTPUT RETURN with a 10Ω resistor.

It is also important to note that when remote sense is used, the voltage across the converter's output terminals (pins 7 and 8) will be higher than the converter's nominal output voltage due to resistive drops along the connecting wires. This higher voltage at the terminals produces a greater voltage stress on the converter's internal components and may cause the converter to fail to deliver the desired output voltage at the low end of the input voltage range at the higher end of the load current and temperature range. Please consult the factory for details.

SYNCHRONIZATION: The MQHR converter's switching frequency can be synchronized to an external frequency source that is in the 500 kHz to 700 kHz range. A pulse train at the desired frequency should be applied to the SYNC IN pin (pin 6) with respect to the INPUT RETURN (pin 2). This pulse train should have a duty cycle in the 20% to 80% range. Its low value should be below 0.8V to be guaranteed to be interpreted as a logic low, and its high value should be above 2.0V to be guaranteed to be interpreted as a logic high. The transition time between the two states should be less than 300ns.

If the MQHR converter is not to be synchronized, the SYNC IN pin should be left open circuit. The converter will then operate in its free-running mode at a frequency of approximately 550 kHz.

If, due to a fault, the SYNC IN pin is held in either a logic low or logic high state continuously, or the SYNC IN frequency is outside the 500-700 kHz range, the MQHR converter will revert to its free-running frequency.

The MQHR converter also has a SYNC OUT pin (pin 5). This output can be used to drive the SYNC IN pins of as many as ten (10) other MQHR converters. The pulse train coming out of SYNC OUT has a duty cycle of 50% and a frequency that matches the switching frequency of the converter with which it is associated. This frequency is either the free-running frequency if there is no valid synchronization signal at the SYNC IN pin, or the synchronization frequency if there is.

The synchronization feature is entirely compatible with that of SynQor's MQFL family of converters.

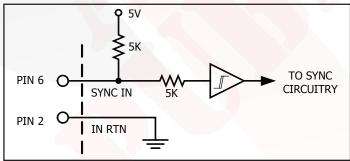


Figure B: Equivalent circuit looking into the SYNC IN pin with respect to the IN RTN (input return) pin.

Figure B shows the equivalent circuit looking into the SYNC IN pin and Figure C shows the equivalent circuit looking into the SYNC OUT pin.

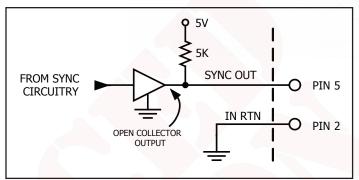


Figure C: Equivalent circuit looking into SYNC OUT pin with respect to the IN RTN (input return) pin.

OUTPUT VOLTAGE TRIM: the TRIM pin (pin 11) can adjust the MQHR converter's output voltage $\pm 10\%$ around its nominal value.

To trim the output voltage above its nominal value, connect an external resistor from the TRIM pin to the –SENSE pin as shown in Figure D. The value of this trim up resistor should be chosen according to the following equation or from Figure E:

Rtrim up(
$$\Omega$$
) = $\frac{7200\Omega*Vnom}{Vout - Vnom} - 36100\Omega$

where:

Vnom = the converter's nominal output voltage, Vout = the desired output voltage (greater than Vnom), and Rtrim up is in Ohms.

As the output voltage is trimmed up, it produces a greater voltage stress on the converter's internal components and may cause the converter to fail to deliver the desired output voltage at the low end of the input voltage range at the higher end of the load current and temperature range. Please consult the factory for details. To trim the output voltage below its nominal value, connect an external resistor (Rtrim down) between the TRIM pin and the +SENSE pin, and another resistor (Rtrim sense) connected between the TRIM pin and the -SENSE pin as shown in Figure D. The values of these trim down resistors should be chosen according to the following equation or from Figure E:

Rtrim down(
$$\Omega$$
) = $\frac{26000\Omega*Vout - 14300\Omega*Vnom}{Vnom - Vout}$ - 71500 Ω

Rtrim sense(Ω) = 1.02 * Rtrim down(Ω)

where:

Vnom = the converter's nominal output voltage, Vout = the desired output voltage (less than Vnom), and

Rtrim down and Rtrim sense are in Ohms. Factory trimmed converters are available by request.

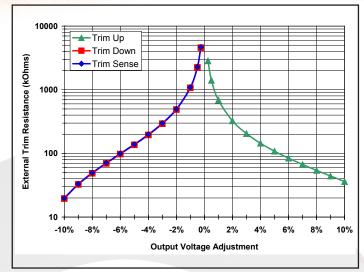


Figure E: Trim up and Trim down as a function of external trim resistance.

INPUT UNDER-VOLTAGE SHUTDOWN: The MQHR converter has an under-voltage shutdown feature that ensures the converter will be off if the input voltage is too low. The input voltage turn-on threshold is higher than the turn-off threshold. In addition, the MQHR converter will not respond to a state of the input voltage unless it has remained in that state for more than about 200µs. This hysteresis and the delay ensure proper operation when the source impedance is high or in a noisy environment.

INPUT OVER-VOLTAGE SHUTDOWN: The MQHR converter also has an over-voltage feature that ensures the converter will be off if the input voltage is too high. It also has a hysteresis and time delay to ensure proper operation.

OUTPUT OVER-VOLTAGE SHUTDOWN: The MQHR converter will shut down if the voltage at its power output pins ever exceeds about 130% of the nominal value. The

shutdown threshold does not change with output trim or sense drops; excessive trim-up or output wiring drops may cause an output over-voltage shutdown event. After a startup inhibit delay, the converter will attempt to restart.

OUTPUT UNDER-VOLTAGE SHUTDOWN: The MQHR converter will also shut down if the voltage at its power output pins ever dips below 60% of the nominal value for more than a few milliseconds. Output voltage reduction due to output current overload (current limit) is the most common trigger for this shutdown. The shutdown threshold does not change with output trim but at only 10%, trimdown should not trigger this event. After a startup inhibit delay, the converter will attempt to restart. This shutdown is disabled during startup.

BACK-DRIVE CURRENT LIMIT: Converters that use MOSFETs as synchronous rectifiers are capable of drawing a negative current from the load if the load is a source of short- or long-term energy. This negative current is referred to as a "back-drive current".

Conditions where back-drive current might occur include paralleled converters that do not employ current sharing. It can also occur when converters having different output voltages are connected together through either explicit or parasitic diodes that, while normally off, become conductive during startup or shutdown. Finally, some loads, such as motors, can return energy to their power rail. Even a load capacitor is a source of back-drive energy for some period of time during a shutdown transient.

To avoid any problems that might arise due to back-drive current, the MQHR converters limit the negative current that the converter can draw from its output terminals. The threshold for this back-drive current limit is placed sufficiently below zero so that the converter may operate properly down to zero load, but its absolute value (see the Electrical Characteristics page) is small compared to the converter's rated output current.

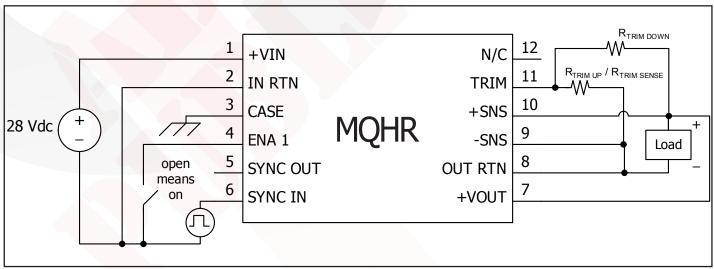


Figure D: Typical connection for output voltage trimming.

CURRENT LIMIT: In the event of excess load, the MQHR converter will quickly reduce its output voltage to keep the load current within safe limits (see Figure 6). If the overload persists for more than 14 milliseconds, the converter will shut off, wait a restart delay, and then automatically attempt to re-start. The timeout is internally implemented with an integrator: counting up whenever current limit is active, and counting down at 1/5th the rate whenever current limit becomes inactive. In this way a series of short-duration overloads will not cause the converter to shut down, while it will shut down in response to sustained overloads.

THERMAL CONSIDERATIONS: Figure 5 shows the suggested Power Derating Curves for this converter as a function of the case temperature and the maximum desired power MOSFET junction temperature. All other components within the converter are cooler than its hottest MOSFET, which at full power is no more than 20°C higher than the case temperature directly below this MOSFET.

The Mil-HDBK-1547A component derating guideline calls for a maximum component temperature of 105°C. Figure 5 therefore has one power derating curve that ensures this limit is maintained. It has been SynQor's extensive experience that reliable long-term converter operation can be achieved with a maximum component temperature of 125°C. In extreme cases, a maximum temperature of 145°C is permissible, but not recommended for long-term operation where high reliability is required. Derating curves for these higher temperature limits are also included in Figure 5. The maximum case temperature at which the converter should be operated is 135°C.

When the converter is mounted on a metal plate, the plate will help to make the converter's case bottom a uniform temperature. How well it does so depends on the thickness of the plate and on the thermal conductance of the interface layer (e.g. thermal grease, thermal pad, etc.) between the case and the plate. Unless this is done very well, it is important not to mistake the plate's temperature for the maximum case temperature. It is easy for them to be as much as 5-10°C different at full power and at high temperatures. It is suggested that a thermocouple be attached directly to the converter's case through a small hole in the plate when investigating how hot the converter is getting. Care must also be made to ensure that there is not a large thermal resistance between the thermocouple and the case due to whatever adhesive might be used to hold the thermocouple in place.

INPUT SYSTEM INSTABILITY: This condition can occur because any dc-dc converter appears incrementally as a negative resistance load. A detailed application note titled "Input System Instability" is available on the SynQor website which provides an understanding of why this instability arises, and shows the preferred solution for correcting it.

Output: 9V

Current: 2.75A

		C-Grade	ES-Grade	HB-Grade
Screening	Consistent with MIL-STD-883F	(specified from 0 °C to +70 °C)	(specified from (-45 °C to +100 °C)	specified from - -55 °C to +125 °C
Element Evaluation		No	Yes	Yes
Internal Visual	*	Yes	Yes	Yes
Temperature Cycle	Method 1010	No	Condition B (-55 °C to +125 °C)	Condition C (-65 °C to +150 °C)
Constant Acceleration	Method 2001 (Y1 Direction)	No	500g	Condition A (5000g)
Burn-in	Method 1015	24 Hrs @ +125 ℃	96 Hrs @ +125 ℃	160 Hrs @ +125 ℃
Final Electrical Test	Method 5005 (Group A)	+25 °C	-45, +25, +100 °C	-55, +25, +125 °C
Mechanical Seal, Thermal, and Coating Process		Full QorSeal	Full QorSeal	Full QorSeal
External Visual	2009	*	Yes	Yes
Construction Process		QorSeal	QorSeal	QorSeal

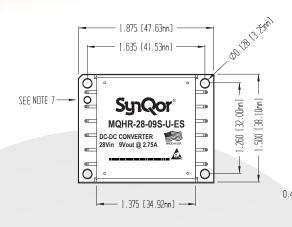
* Per IPC-A-610 Class 3

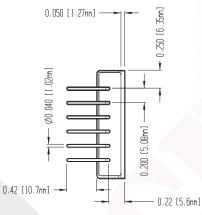
MilQor converters and filters are offered in three variations of environmental stress screening options. All MilQor converters use SynQor's proprietary QorSeal™ Hi-Rel assembly process that includes a Parylene-C coating of the circuit, a high performance thermal compound filler, and a nickel barrier gold plated aluminum case. Each successively higher grade has more stringent mechanical and electrical testing, as well as a longer burn-in cycle. The ES-and HB-Grades are also constructed of components that have been procured through an element evaluation process that pre-qualifies each new batch of devices.

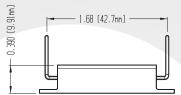


MQHR-28-09S Output: 9V

Current: 2.75A



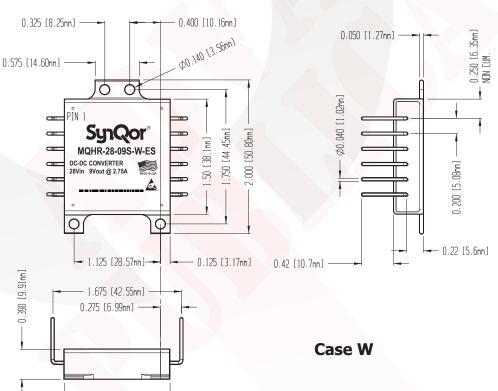




Case U

PIN DESIGNATIONS

Pin #	Function
1	Positive input
2	Input return
3	Case
4	Enable 1
5	Sync output
6	Sync input
7	Positive output
8	Output return
9	- Sense
10	+ Sense
11	Trim
12	No connection



NOTES

- Case: Aluminum with gold over nickel plate finish for the C, ES, and HB Grade products.
- 2) Pins: Diameter: 0.040" (1.02mm)

 Material: Copper
 Finish: Copper alloy with Gol
 - Finish: Copper alloy with Gold over Nickel plating, followed by Sn/Pb solder dip
- 3) All dimensions in inches (mm)
- 4) Tolerances: a) x.xx +/-0.02 in. (x.x +/-0.5mm) b) x.xxx +/-0.010 in.
 - b) x.xxx +/-0.010 in. (x.xx +/-0.25mm)
- 5) Weight: 1.6 oz (45.4 g) typical
- 6) Workmanship: Meets or exceeds IPC-A-610 Class III
- 7) Pin 1 identification hole, not intended for mounting
- 8) Baseplate flatness tolerance is 0.004" (.10mm) TIR for surface

- 1.375 [34.92mm]

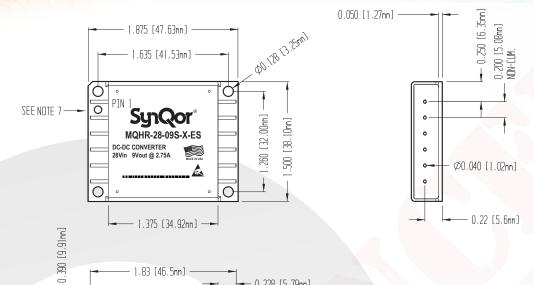


1.83 [46.5mm]

MQHR-28-09S

Output: 9V

Current: 2.75A

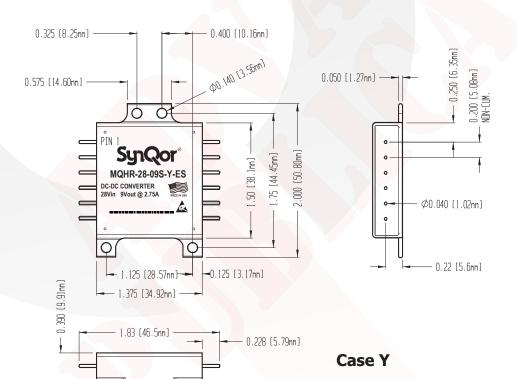


0.228 [5.79mm]

PIN DESIGNATIONS

Function
Positive input
Input return
Case
Enable 1
Sync output
Sync input
Positive output
Output return
- Sense
+ Sense
Trim
No connection

Case X



NOTES

- Case: Aluminum with gold over nickel plate finish for the C, ES, and HB Grade products.
- Diameter: 0.040" (1.02mm) 2) Pins:
 - Material: Copper Finish: Copper alloy with Gold over Nickel plating, followed by Sn/Pb solder dip
- All dimensions in inches (mm) 3)
- 4) a) x.xx + /-0.02 in. Tolerances:

(x.x + /-0.5mm)b) x.xxx + /-0.010 in. (x.xx + /-0.25mm)

- 5) Weight: 1.6 oz (45.4 g) typical
- 6) Workmanship: Meets or exceeds IPC-A-610 Class III
- 7) Pin 1 identification hole, not intended for mounting
- Baseplate flatness tolerance is 0.004" (.10mm) TIR for surface



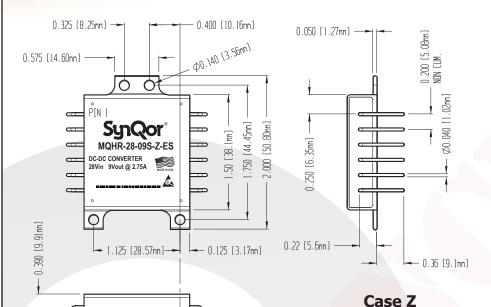
0.28 [7.1mm] - 1.375 [34.92mm]

1.68 [42.7nm]

MQHR-28-09S

Output: 9V

Current: 2.75A



PIN DESIGNATIONS

Pin #	Function
1	Positive input
2	Input return
3	Case
4	Enable 1
5	Sync output
6	Sync input
7	Positive output
8	Output return
9	- Sense
10	+ Sense
11	Trim
12	No connection

NOTES

- 1) Case: Aluminum with gold over nickel plate finish for the C,
 - ES, and HB Grade products.
- 2) Pins: Diameter: 0.040" (1.02mm)
 - Material: Copper
 - Finish: Copper alloy with Gold over Nickel plating, followed by Sn/Pb solder dip
- 3) All dimensions in inches (mm)
- 4) Tolerances: a) x.xx +/-0.02 in.
 - (x.x +/-0.5mm) b) x.xxx +/-0.010 in. (x.xx +/-0.25mm)
- 5) Weight: 1.6 oz (45.4 g) typical
- 6) Workmanship: Meets or exceeds IPC-A-610 Class III
- 7) Pin 1 identification hole, not intended for mounting
- 8) Baseplate flatness tolerance is 0.004" (.10mm) TIR for surface

Output: 9V

Current: 2.75A

MilQor Converter FAMILY MATRIX

The tables below show the array of MilQor converters available. When ordering SynQor converters, please ensure that you use the complete part number according to the table in the last page. Contact the factory for other requirements.

					Sino	gle Ou	tput					D	ual Out	put †
Full Size	1.5V	1.8V	2.5V	3.3V	5V	6V	7.5V	9V	12V	15V	28V	5V	12V	15V
	(1R5S)	(1R8S)	(2R5S)	(3R3S)	(05S)	(06S)	(7R5S)	(09S)	(12S)	(15S)	(28S)	(05D) (12D)	(15D)
MQFL-28 16-40Vin Cont. 16-50Vin 1s Trans.* Absolute Max Vin = 60V	40A	40A	40A	30A	24A	20A	16A	13A	10A	8A	4A	24A Tota		8A Total
MQFL-28E 16-70Vin Cont. 16-80Vin 1s Trans.* Absolute Max Vin =100V	40A	40A	40A	30A	24A	20A	16A	13A	10A	8A	4A	24A Tota		8A Total
MQFL-28V 16-40Vin Cont. 5.5-50Vin 1s Trans.* Absolute Max Vin = 60V	40A	40A	40A	30A	20A	17A	13A	11A	8A	6.5A	3.3A			
MQFL-28VE 16-70Vin Cont. 5.5-80Vin 1s Trans.* Absolute Max Vin = 100V	40A	40A	40A	30A	20A	17A	13A	11A	8A	6.5A	3.3A			
MQFL-270 155-400Vin Cont. 155-475Vin 1s Trans.* Absolute Max Vin = 550V	40A	40A	40A	30A	24A	20A	16A	13A	10A	8A	4A	24A Tota	10A I Total	8A Total
MQFL-270L 65-350Vin Cont. 65-475Vin 1s Trans.* Absolute Max Vin = 550V	40A	40A	30A	22A	15A	12A	10A	8A	6A	5A	2.7A	15A Tota		5A Total
						gle Ou							ual Out	_
Half Size	1.5V (1R5S)	1.8V (1R8S)	2.5V (2R5S)	3.3V (3R3S)	5V (05S)	6V (06S)	7.5V (7R5S)	9V (09S)	12V (12S)	15V (15S)	28V (28S)	5V (05D	12V (12D)	15V (15D)
MQHL-28 16-40Vin Cont. 16-50Vin 1s Trans.* Absolute Max Vin = 60V	20A	20A	20A	15A	10A	8A	6.6A	5.5A	4A	3.3A	1.8A	10A Tota	4A	3.3A Total
MQHL-28E 16-70Vin Cont. 16-80Vin 1s Trans.* Absolute Max Vin =100V	20A	20A	20A	15A	10A	8A	6.6A	5.5A	4A	3.3A	1.8A	10A Tota	4A I Total	3.3A Total
MQHR-28 16-40Vin Cont. 16-50Vin 1s Trans.* Absolute Max Vin = 60V	10A	10A	10A	7.5A	5A	4A	3.3A	2.75A	2A	1.65A	0.9A	5A Tota	2A I Total	1.65A Total
MQHR-28E 16-70Vin Cont. 16-80Vin 1s Trans.* Absolute Max Vin = 100V	10A	10A	10A	7.5A	5A	4A	3.3A	2.75A	2A	1.65A	0.9A	5A Tota	2A I Total	1.65A Total

Check with factory for availability.

^{†80%} of total output current available on any one output.

^{*}Converters may be operated at the highest transient input voltage, but some component electrical and thermal stresses would be beyond MIL-HDBK-1547A guidelines.

PART NUMBERING SYSTEM

The part numbering system for SynQor's MilQor DC-DC converters follows the format shown in the table below.

Not all combinations make valid part numbers, please contact SynQor for availability. See the Product Summary web page for more options.

Example: MQHR-28-09S-Y-ES

Model	Input	Output V	/oltage(s)	Package Outline/	Scrooning		
Name	Voltage Range	Single Output	Dual Output	Pin Configuration	Screening Grade		
MQFL MQHL MQHR	28 28E 28V 28VE 270 270L	1R5S 1R8S 2R5S 3R3S 05S 06S 7R5S 09S 12S 15S 28S	05D 12D 15D	U X Y W Z	C ES HB		

APPLICATION NOTES

A variety of application notes and technical white papers can be downloaded in pdf format from the SynQor website.

PATENTS

SynQor holds the following U.S. patents, one or more of which apply to each product listed in this document. Additional patent applications may be pending or filed in the future.

5,999,417	6,222,742	6,545,890	6,577,109	6,594,159	6,731,520
6,894,468	6,896,526	6,927,987	7,050,309	7,072,190	7,085,146
7,119,524	7,269,034	7,272,021	7,272,023	7,558,083	7,564,702
7,765,687	7,787,261	8,023,290	8,149,597		

Contact SynQor for further information and to order:

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Web: www.synqor.com Address: 155 Swanson Road

Boxborough, MA 01719

USA

Warranty

SynQor offers a two (2) year limited warranty. Complete warranty information is listed on our website or is available upon request from SynQor.

Information furnished by SynQor is believed to be accurate and reliable. However, no responsibility is assumed by SynQor for its use, nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of SynQor.